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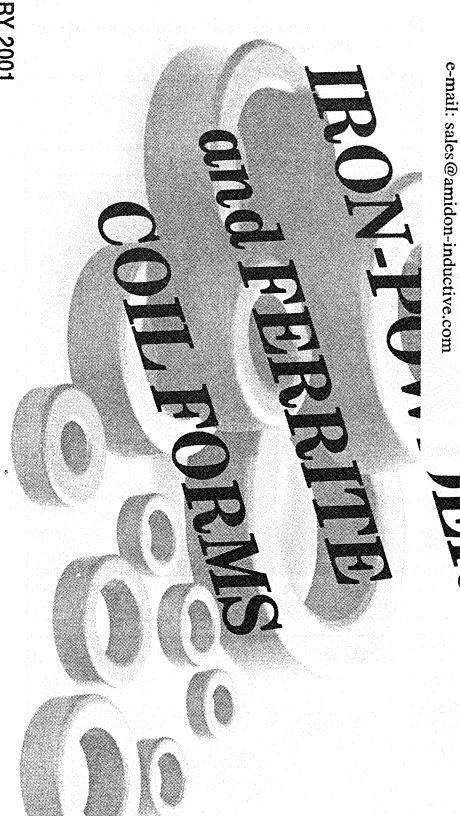
Amidon Inductive Components

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IRON POWDER CORES and MATERIALS

There are two basic material groups of Iron Powders cores. The Carbonyl Irons and the Hydrogen Reduced Irons.

The Carbonyl Irons are especially noted for their excellent stability over a wide range of flux levels and temperatures. Their permeability range is from one to 35 mu they can offer excellent 'Q' factors up to more than 200 MHz. They are widely used for broadband inductors, especially in higher power applications.

The Hydrogen Reduced Irons have higher permeabilities up to 90 mu. This group will have somewhat lower 'Q' and they are mainly used for EMI filters, low frequency chokes, and input and output filters for switched mode power supplies.

Toroidal cores are self shielding and it is not necessary to isolate or shield a them to prevent cross talk or feedback. Each Iron Powder core an $A_{\rm L}$ value listed in uh/100 turns and can be found elsewhere in this paper. Turns for a desired inductance may be calculated by using the given $A_{\rm L}$ value and the formula below.

Turns Formula Key to part number

$$Turns = 100 \sqrt{\frac{\text{desired L (uh)}}{\text{AL (uh/100 t)}}} \frac{\text{T}}{\text{toroid outer-diameter}} - \frac{6}{\text{material}}$$

MATERIAL #0 (u=1) Most commonly used for frequencies above 200 MHz. Available in toroidal form only. Inductance vs. turns may vary according to winding technique.

MATERIAL #1 (u=20) A Carbonyl 'C' material very similar to material #3 except that it has higher volume resistivity and better stability. Available in both toroidal form and shielded coil form.

MATERIAL #2 (u=10) A carbonyl 'E' iron powder material having high volume resistivity and offers high 'Q' for the 1 to 30 MHz. frequency range. Available in both toroidal and shielded coil forms.

MATERIAL #3 (u=35) A carbonyl 'HP' material having excellent stability and good 'Q' for the lower frequencies from 50 KHz. to 500 KHz. Available in toroidal and shielded coil forms.

MATERIAL #6 (u=8) A carbonyl 'SF' material very similar to #2 material but has an improved 'Q' for frequencies 20 MHz to 50 MHz. Available in toroidal core form and shielded coil form.

MATERIAL #10 (u=6) A powdered iron 'W' material. Offers good 'Q' and high stability for frequencies 40 Mhz to 100 MHz. Available in both toroidal form and shielded coil form.

MATERIAL #12 (u=3) A Synthetic Oxide material which will provide good 'Q'and moderate stability for frequencies 50 Mhz to 100 MHz. Available in toroidal form only.

MATERIAL #15 (u=25) A carbonyl 'GS6' material. Has excellent stability and good 'Q'. A good choice for commercial broadcast frequencies where good 'Q' and stability are essential. Available in toroidal form only.

MATERIAL #17 (u=3) This is a carbonyl material very similar to that of the of the #12 material. It has greater temperature stability but the cost of somewhat lower `Q'. Available in all shielded coil forms. In the toroidal form, only sizes T-12 through T-50 are available.

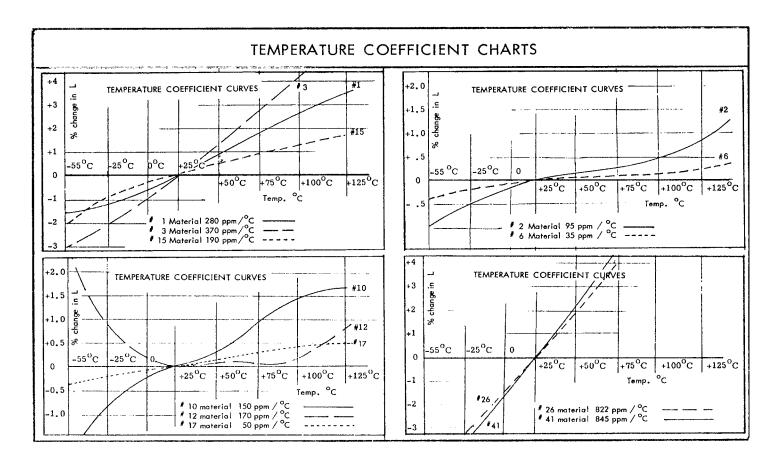
MATERIAL #26 (u=75) A Hydrogen Reduced material. Has highest permeability of all of the iron powder materials. Used for EMI filters and DC chokes. The #26 material is very similar to the older #41 material but provides an extended frequency range. Available in all toroidal core sizes.

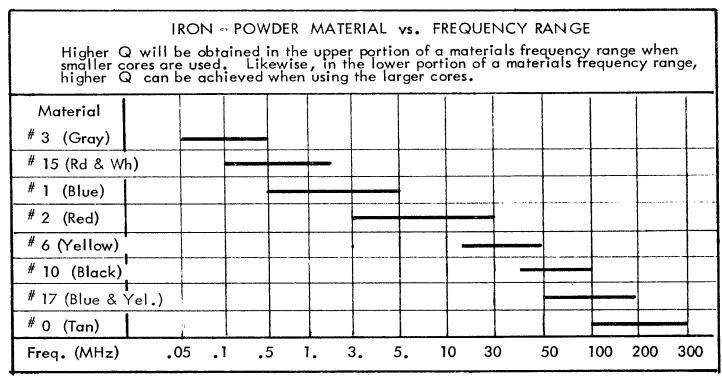
IRON POWDER TOROIDAL CORES

			F	hys:	ica1	Di	mens	ions				
Core Size \/	Outer diam. (in)	Inner diam. (in)	Height	Mean lgth. (cm)	Cross sect. (cm ²)		Core Size	Outer diam (in)	Inner diam (in)	Height	Mean lgth. (cm)	Cross sect. (cm ²)
T- 12	.125	.062	.050	0.75	.010		T-130	1.30	.78	.437	8.29	0.73
T- 16	.160	.078	.060	0.95	.016		T-157	1.57	.95	.570	10.05	1.14
T- 20	.200	.088	.070	1.15	.025		T-184	1.84	.95	.710	11.12	2.04
T- 25	.250	.120	.096	1.50	.042		T-200	2.00	1.25	.550	12.97	1.33
T- 30	.307	.151	.128	1.83	.065		T-200A	2.00	1.25	1.000	12.97	2.42
T- 37	.375	.205	.128	2.32	.070		T-225	2.25	1.40	.550	14.56	1.50
T- 44	.440	.229	.159	2.67	.107		T-225A	2.25	1.40	1.000	14.56	2.73
T- 50	.500	.300	.190	3.20	.121		T-300	3.00	1.92	.500	19.83	1.81
T- 68	.690	.370	.190	4.24	.196		T-300A	3.00	1.92	1.000	19.83	3.58
T- 80	.795	.495	.250	5.15	.242		T-400	4.00	2.25	.650	24.93	3.66
T- 94	.942	.560	.312	6.00	.385		T-400A	4.00	2.25	1.000	24.93	7.43
T-106	1.060	.570	.437	6.50	.690		T-500	5.20	3.08	.800	33.16	5.46

			alues complete p	(u art number,		OO tu	irns) Core Size n	umber.	
Core Size \/ Mhz >	26 Mix Yel-Wh u = 75 to 1.0	3 Mix Gray u = 35 .055	15 Mix Rd-Wh u = 25	1 Mix Blue u = 20 .5 - 5.	2 Mix Red u = 10 1 - 30	6 Mix Yellow u = 8 3 - 50	10 Mix Black u = 6 5 - 100	12/17 Mix Grn-Wh u = 3.5 20-200	0 Mix Tan u = 1 50-300
T- 12-	na	60	50	48	20	17	12	7.0	3.0
T- 16-	na	61	55	44	22	19	13	8 0	3.0
T- 20-	na	90	65	52	27	22	16	10.0	3.5
T- 25-	na	100	100	70	34	27	19	12.0	4.5
T- 30-	325	140	93	85	43	36	25	16.0	6.0
T- 37-	275	120	90	80	40	30	25	15.0	4.9
T- 44-	360	180	160	105	52	42	33	19.0	6.5
T- 50-	320	175	135	100	49	40	31	18.0	6.4
T- 68-	420	195	180	115	57	47	32	21.0	7.5
T- 80-	450	180	170	115	55	45	32	22.0	8.5
T- 94-	590	248	200	160	84	70	58	32.0	10.6
T-106-	900	450	345	325	135	116	na	na	19.0
T-130-	785	350	250	200	110	96	na	na	15.0
T-157-	970	420	360	320	140	115	na	na	na
T-184-	1640	720	na	500	240	na	na	na	na
T-200-	895	425	na	250	120	100	na	na	na
T-200A-	1550	na	na	na	218	180	na	na	na
T-225-	950	424	na	na	120	100	na	na	na
T-225A-	1600	na	na	na	215	na	na	na	na
T-300-	825	na	na	na	115	na	na	na	na
T-300A-	1600	na	na	na	228	na	na	na	na
T-400-	1320	na	na	na	185	na	na	na	na
T-400A-	2600	na	na	na	360	na	na	na	na
T-500-	1460	na	na	na	207	na	na	na	na
na - not a	vailable.		Note: #12	2 material	will event	ually be s	superseded b	y the #17 ma	aterial.

Iron Powder Toroidal Cores



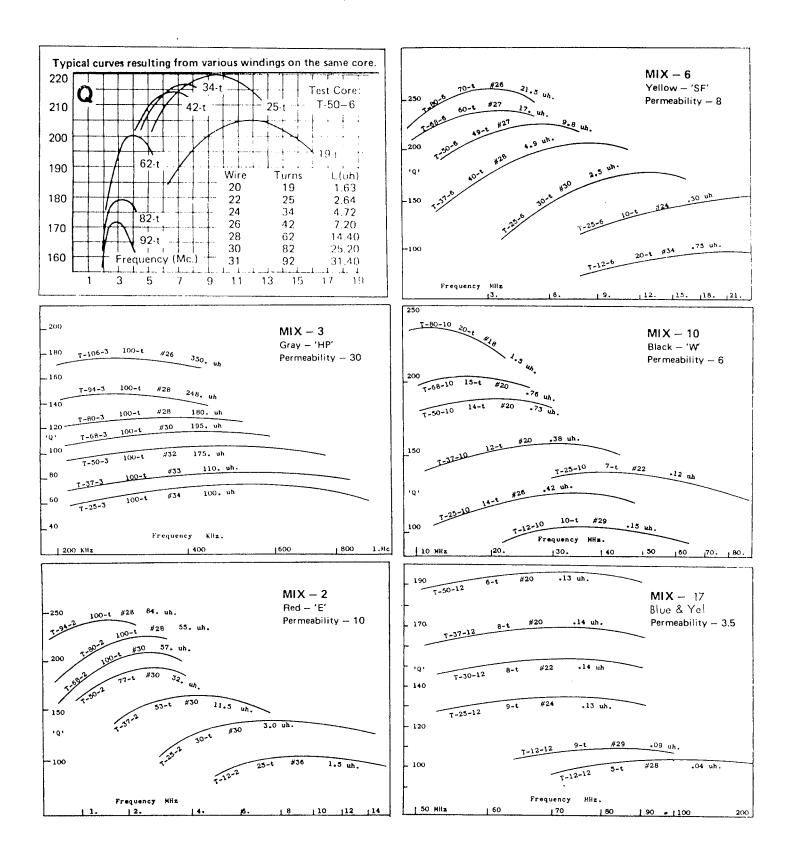


	+	Coppe	r Wire	Table	e	
Wire size AWG	Diameter in inches (enamel)	Circular mil area	Turns per linear inch	Turns per sq.cm	Contineous duty current (amps) single wire,open air	Contineous duty,(amps) conduit or in wire bundles
8 10 12	.1285 .1019 .0808	16510 10380 6530	7.6 10.7 12.0	13.8 21.7	73 55 41	46 33 23
14 16 18	.0640 .0508 .0403	4107 2583 1624	15.0 18.9 23.6	34.1 61.2 79.1	32 22 16	17 13 10
20 22 24	.0319 .0253 .0201	1022 642 404	29.4 37.0 46.3	124.0 186.0 294.0	11 	7.5 5.0
26 28 30	.0159 .0126 .0100	254 160 101	58.0 72.7 90.5	465.0 728.0 1085.0	 	
32 34 36	.0079 .0063 .0050	63 40 25	113.0 141.0 175.0	1628.0 2480.0 3876.0	 	
38 40	.0039	16 10	224.0 382.0	5736.0 10077.0		

Iron	Ро	wd qqA		C nate			Siz f tur				ns ingle	& laye:	Wind r wind		Si	ze
Wire Sz.	10	12	14	16	18	20	22	24	26	28	30	32	34	36	38	40
Core\/No																
T-12	0	0	0	1	1	1	2	4	5	8	11	15	21	29	37	47
T-16	0	0	1	1	1	3	3	5	8	11	16	21	29	38	49	63
T-20	0	1	1	1	3	4	5	6	9	14	18	25	33	43	56	72
T-25	1	1	1	3	4	5	7	11	15	21	28	37	48	62	79	101
T-30	1	1	3	4	5	7	11	15	21	28	37	48	62	78	101	129
T-37	1	3	5	7	9	12	17	23	31	41	53	67	87	110	140	177
T-44	3	5	6	7	10	15	20	27	35	46	60	76	97	124	157	199
T-50	5	6	8	11	16	21	28	37	49	63	81	103	131	166	210	265
T-68	7	9	12	15	21	28	36	47	61	79	101	127	162	205	257	325
T-80	8	12	17	23	30	39	51	66	84	108	137	172	219	276	347	438
T-94	10	14	20	27	35	45	58	75	96	123	156	195	248	313	393	496
T-106	10	14	20	27	35	45	58	75	96	123	156	195	248	313	393	496
T-130	17	23	30	40	51	66	83	107	137	173	220	275	348	439	550	693
T-157	22	29	38	50	64	82	104	132	168	213	270	336	426	536	672	846
T-184	22	29	38	50	64	82	104	132	168	213	270	336	426	536	672	846
T-200	31	41	53	68	86	109	139	176	223	282	357	445	562	707	886	1115
T-225	36	46	60	77	98	123	156	198	250	317	400	499	631	793	993	1250
T-300	52	66	85	108	137	172	217	274	347	438	553	688	870	1093	1368	1721
T-400 T-520	61 86	79 110		127 160			255 349	322 443	407 559	513 706	648 889	806 1105		1278 1753		

IRON POWDER TOROIDAL CORES

'Q' CURVES



AC LINE FILTERS and DC CHOKES (# 26 Material)

High 'Q' inductors are no longer required for energy storage applications, in fact low 'Q' actually helps damp high frequency oscillations. The #26 Iron Powder material is ideally suited for these applications since it combines low 'Q', good frequency response, and high energy capabilities.

Energy storage, expressed in microjuoles, is calculated by multiplying one-half the inductance in uh times the current in amperes squared. The amount of energy that can be stored in a given inductor is limited either by saturation of the core material or temperature rise of the wound unit, resulting in copper loss and/or core loss.

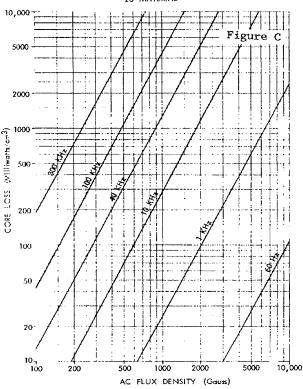
In typical DC chokes, the AC ripple flux is normally small in comparison to the DC component. Since the DC flux does not generate core loss, our primary concern becomes saturation and copper loss. The DC saturation characteristics of the #26 material are shown in Fig. A.

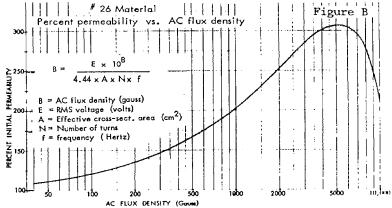
In 60 Hz. line filter applications, the high frequency to be filtered falls into two categories: (1) Common-mode noise and (2) Differential-mode noise. The common-mode noise is in relation to earth ground and is common to both lines. Differential mode noise is the noise between the two lines.

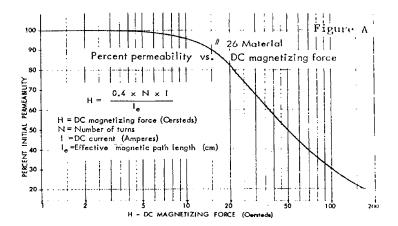
Iron Powder cores are not recommended for Common-mode noise filters. They are usually constructed on a high permeability FERRITE core with a bifilar winding. This allows the 60 Hz. flux generated by each line to cancel within the core.

The Iron Powder #26 material is ideal for Differential-mode filters, since the core must be able to support a significant amount of 60Hz. flux without saturating. The AC saturation characteristics of the #26 material (Fig. B) and core loss information (Fig. C) can be seen below. Notice how the permeability initially increases with AC excitation. This effect allows greater energy storage in 60 Hz. applications.

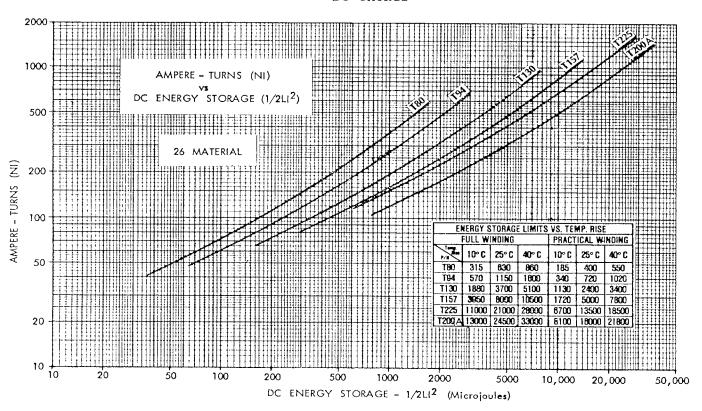






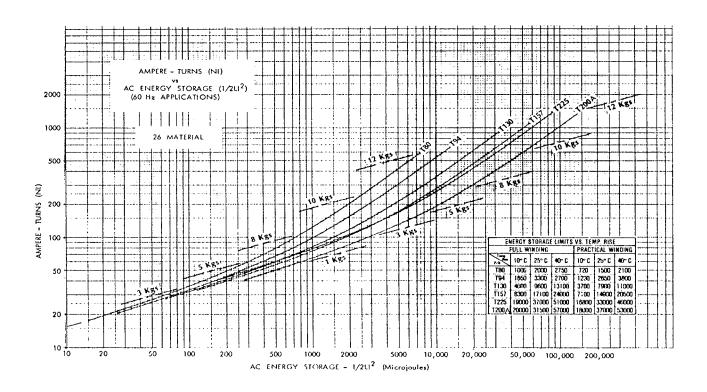


Iron Powder Toroidal Cores DC Chokes



	DC C	hoke	Appl	icati	ons	(single lay	er winding)	
DC Amps > Wire size >	1 Amp	2 Amps	4 Amps	6 Amps	10 Amps	15 Amps	20 Amps	30 Amps
	28 AWG	24 AWG	21 AWG	19 AWG	16 AWG	14 AWG	12 AWG	10 AWG
Part.\/No								
T-31-26*	35 uh	13.5 uh	4.0 uh	1.8 uh	.8 uh	.38 uh	.16 uh	.012 uh
	41 turns	27 turns	15 turns	10 turns	7 turns	5 turns	3 turns	1 turn
T-50-26*	92 uh	29.0 uh	11.3 uh	5.5 uh	2.1 uh	1.1 uh	.59 uh	.36 uh
	63 turns	37 turns	25 turns	18 turns	11 turns	8 turns	6 turns	5 turns
T-80-26	380 uh	130 uh	51.3 uh	27.8 uh	11.2 uh	5.7 uh	3 uh	1.3 uh
	108 turns	s 66 turns	45 turns	35 turns	23 turns	17 turns	12 turns	8 turns
T 94-26	650 uh	220 uh	87.5 uh	47.2 uh	20.0 uh	10.2 uh	5.3 uh	2.6 uh
	123 turn	75 turns	52 turns	40 turns	27 turns	20 turns	14 turns	10 turns
T-130-26	1660 uh	575 uh	231 uh	127 uh	55.0 uh	28.0 uh	16.5 uh	10.4 uh
	173 turn	s 107 turns	75 turns	58 turns	40 turns	30 turns	23 turns	17 turns
T 157~26	3200 uh	1100 uh	438 uh	244 uh	106 uh	55.6 uh	32 uh	16.4 uh
	213 turn	s 122 turns	93 turns	73 turns	50 turns	38 turns	29 turns	22 turns
T-184-26*	5600 uh	1950 uh	788 uh	439 uh	190 uh	99.6 uh	57.5 uh	29.3 uh
	213 turn	s 122 turns	93 turns	73 turns	50 turns	38 turns	29 turns	22 turns
T 225-26	8600 uh	2300 uh	938 uh	528 uh	230 uh	127 uh	72.5 uh	40 uh
	317 turn	s 198 turns	139 turns	110 turns	77 turns	60 turns	46 turns	36 turns
T : 300A=26 ²	22.4 mh	7850 uh	3120 uh	1750 uh	760 uh	418 uh	250 uh	129 uh
	435 turn	s 272 turns	190 turns	151 turns	105 turns	82 turns	63 turns	44 turns
T 400A 26*	51.0 mh	17.5 mh	7120 uh	4000 uh	1760 uh	951 uh	550 uh	293 uh
	507 turn	s 317 turns	223 turns	176 turns	122 turns	95 turns	73 turns	57 turns
Not o:	* Size n	ot shown on	above curve	chart.	%% Wire siz	e based on M	lax. Temp. r	ise 40E0 C.

Iron Powder Toroidal Cores AC Line Filters



60 Hz. A C	C Lin	e Fi	lter	Appli	icati	ons (s	ingle layer	winding)
AC Amps >	1 Amp	2 Amps	4 Amps	6 Amps	10 Amps	15 Amps	20 Amps	30 Amps
Wire size >	28 AWG	24 AWG	21 AWG	19 AWG	15 AWG	13 AWG	11 AWG	9 AWG
Part\/No								
T-37 -26%	130 uh	50.0 uh	15 uh	6.7 uh	2.4 uh	1.1 uh	.60 uh	.07 uh
	41 turns	27 turns	15 turns	10 turns	6 turns	4 turns	3 turns	1 turn
T-50 -26*	460 uh	150 uh	58.8 uh	26.1 uh	9.4 uh	4.2 uh	2.4 uh	1.0 uh
	63 turns	37 turns	25 turns	17 turns	10 turns	7 turns	5 turns	3 turns
T-80 -26	1600 uh	550 uh	213 uh	94.4 uh	34.0 uh	15.1 uh	8.5 uh	3.8 uh
	108 turns	66 turns	45 turns	30 turns	18 turns	12 turns	9 turns	6 turns
T-94 -26	2899 uh	950 uh	375 uh	156 uh	56.0 uh	24.9 uh	14 uh	6.2 uh
	123 tunrs	75 turns	52 turns	33 turns	20 turns	13 turns	10 turns	7 turns
T-130 -26	7200 uhh	2500 uh	1000 uh	444 uh	160 uh	71.1 uh	40 uh	17.8 uh
	173 turns	107 turns	75 turns	50 turns	30 turns	20 turns	15 turns	10 turns
T-157 -26	13.6 mh	4650 uh	1810 uh	806 uh	290 uh	129 uhh	72.5 uh	32.2 uh
	213 turns	139 turns	93 turns	62 turns	37 turns	25 turns	18 turns	12 turns
T-184 -26*	22 mh	7750 uh	3130 uh	1390 uh	500 uh	222 uh	125 uh	56.6 uh
	213 turns	132 turns	93 turns	62 turns	37 turns	25 turns	18 turns	12 turns
T-225 -26	26 mh	9000 uh	3500 uh	1940 uh	700 uh	311 uh	175 uh	77.8 uh
	317 turns	198 turns	139 turns	110 turns	66 turns	44 turns	33 turns	22 turns
T-300A -26*	84 mh	29 mh	11.2 mh	6390 uh	2360 uh	1240 uh	750 uh	356 uh
	435 turns	272 turns	190 turns	151 turns	93 turns	72 turns	56 turns	40 turns
T-400A -26%	180 mh	61 mh	25.6 mh	14.2 mh	5300 uh	2800 uh	1650 uh	800 uh
	507 turns	317 turns	223 turns	176 turns	108 turns	83 turns	65 turns	46 turns
Note:	* Size not	shown on a	bove curve	chart. '6	Wire size N	based on Max	. Temp. rise	e 40 ⁰ C.

POWER CONSIDERATIONS (IP and Ferrite)

How large a core is needed to handle a certain amount of power? This is a question often asked, but unfortunately there is no simple answer.

There are several factors involved such as: cross sect. area, turns count, material, and of course the variables of applied voltage and operating frequency.

Overheating of the coil will usually take place long before saturation in most applications above 100 KHz. Now the question becomes 'How large a core must I have to prevent overheating at a given frequency and power level'? Operating frequency is one of the most important factors concerning power capability above 1 MHz. A core working well at 2 MHz may burn up at 30 MHz. with the same drive.

Overheating can be caused by both wire and core material losses. Wire heating is affected by both DC and AC currents, while core heating is affected only by the AC content of the signal. With a normal sinewave signal above 100 KHz, both the Iron Powder and Ferrite type cores will first be affected by overheating caused by core losses, rather than by saturation.

The following extrapolated AC flux density limits can be used as guide-lines for BOTH the Iron Powder and Ferrite cores to avoid excessive heating. Figures may vary slightly according to material used.

Frequency: 100 KHz 1 MHz 7 MHz 14 MHz. 21 MHz 28 MHz. AC Flux Density: 500 gauss 150 gauss 57 gauss 42 gauss 36 gauss 30 gauss

Iron Powder cores (low permeability) are superior to the Ferrite material cores for high power inductors for this reason: Fewer turns will be required by the Ferrite type core for a given inductance. When the same voltage drop is applied across a decreased number of turns, the flux density will increase accordingly. To prevent the flux density from increasing when fewer turns are used, the flux drive will have to be decreased.

Either core material can be used for transformer applications, but both will require 'trade-offs'. Ferrite cores will require fewer turns and will couple better, whereas the Iron Powder cores will require more turns and not couple as well but will tolerate more power and are more stable.

The equation for $B_{\rm max}$ and a sample calculation are shown below: The sample calculation is based on a frequency of 7 MHz, a peak voltage of 25 volts, a primary winding of 15 turns, and a cross-sect. area of 0.133 cm² (as per data sheet). From above guide-lines, $B_{\rm max}$ at 7 MHz should be not more than 57 gauss,

$$B_{\text{max}} = \frac{E_{\text{pk}} \times 10^2}{4.44 \text{ A}_{\text{e}} \text{ N f}} = \frac{25 \times 100}{4.44 \times 0.133 \times 15 \times 7} = 40 \text{ gauss} \\ \begin{bmatrix} E_{\text{pk}} \\ A_{\text{e}} \\ N \\ = \text{ cross-sect. area (cm}^2) \\ N \\ = \text{ number of wire turns} \\ f \\ = \text{ frequency (MHz)} \\ \end{bmatrix}$$

This hypothetical toroid core will have a flux density of $40~{\rm gauss}$. This is well within the above guidelines to prevent overheating.

Core saturation is affected by both AC and DC signals. Saturation will decrease the permeability of the core causing it to have impaired performance or to become inoperative. The safe operating total flux density for most Ferrites is typically 2000 gauss, while the Iron Powders can tolerate up to 5000 gauss.

Both wire heating and magnetic action within the core will contribute to the temperature rise of the coil. This can be calculated with the equation below:

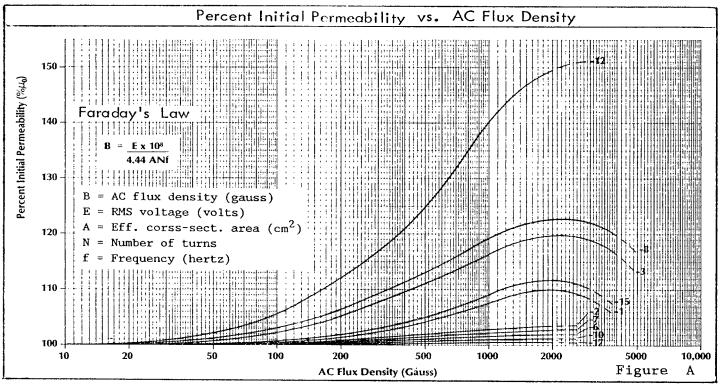
Temperature rise
$${}^{\circ}C = \frac{Power dissipation in milliwatts}{Surface area (cm2)}$$
.833

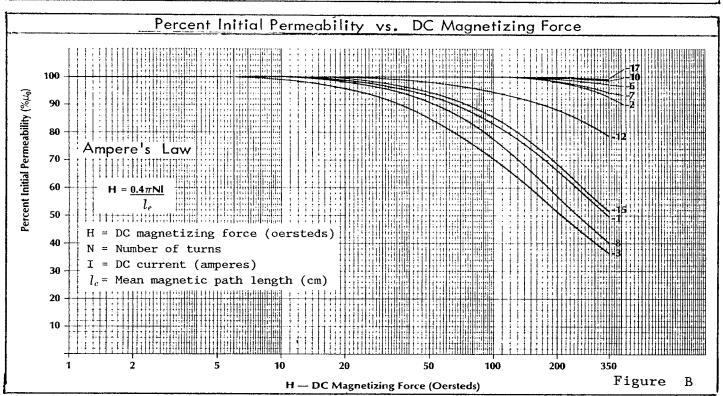
If the operating temperature (ambient temperature + temperature rise) exceeds 100° C when used intermittently, or more that 75° C if used continuously, a larger size core or/and a heavier gauge wire should be selected.

Iron Powder Materials SATURATION and FLUX DENSITY

Factors affecting power capability will vary with operating conditions. Core losses are lower at lc frequencies and low power levels, but increase rapidly as either is increased.

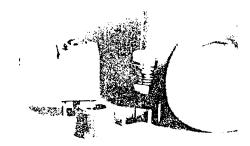
Core losses can create overheating, which in turn will affect the saturation point. Maximum flu density can be calculated with the Faraday Law and Amperes Law, both of which are shown below:





Iron Powder Shielded Coil Forms Slug tuning

L-33 Shielded Coil Form



Sub miniature size
Slug tuning
Copper shield can, tin plated
Easy to wind
Good 'Q'

Frequency range: 0.2 to 200 MHz Inductance range: .02 to 300 uh

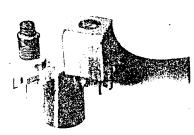
Terminal Spacing	.40 Shiel
.15	Cup L.25 Core

Part number	Frequency	A _L (uh/100t) at max L	L ratio max to min	Typico Wire	al Wind Turns	ling (mid	-freq.) Qmax
L-33-1	0.5 - 1.1	76	1.7 - 1	3/44	75	42.5	80
L-33-2	1.0 - 10.0	68	1.5 - 1	9/44	40	10.9	90
L-33-3	0.2 - 0.6	80	1.8 - 1	3/44	150	180	70
L-33-6	9.0 - 40.0	60	1.5 - 1	26	7	0.36	100
L-33-10	30.0 - 80.0	54	1.4 - 1	26	5	0.18	120
L-33-17	70.0 - 200.0	48	1.3 - 1	26	3	0.08	130



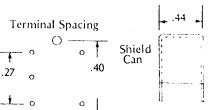
Thread Core

L-43 Shielded Coil Form



Miniature in size
Slug tuning
Copper shield can, tin plated
Easy to wind
Good 'Q'

Frequency range: 0.2 to 200 MHz | 0.2 Inductance range: .02 to 700 uh



Part	Frequency	A _L (uh/100t)	L ratio		Typica	l Windi	ng (mid-	-freq.)
number	range (MHz)	at max L	max to min		Wire	Turns	L(uh)	Qmax
L-43-1	0.5 - 1.1	115	1.6 - 1		5/44	149	230	110
L-43-2	1.0 - 10.0	98	1.6 - 1		9/44	21	4.0	120
L-43-3 L-43-6	0.2 - 0.6 9.0 - 40.0	133 85	1.8 - 1 1.4 - 1		3/44	223 6	600 0.30	90 130
L-43-10	30.0 - 80.0	7 2	1.3 - 1		24	5	0.14	150
L-43-17	70.0 - 200.0	56	1.2 - 1		22	3	0.05	200

Cup Core 32

Threaded Core .25

Base .22

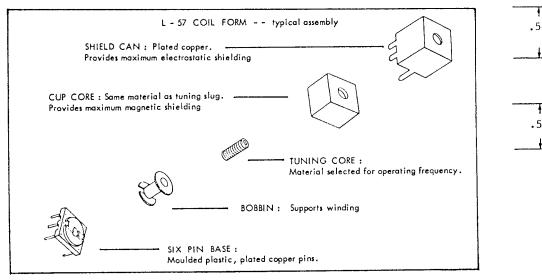
Most efficient when tuning slug is set at maximum L. For tuning flexibility, calculate so that slug will be about 90% maximum L when at operating frequency.

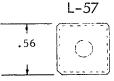
Turns =
$$100 \sqrt{\frac{\text{desired L (uh)}}{90\% \text{ A_L value (uh/100 t)}}}$$

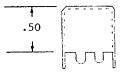
Iron Powder Shielded Coil Forms Slug tuning

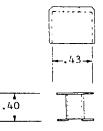
L-57 Shielded Coil Form

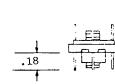
Part Number	Frequency Range	${ m A_L}$ (uh/100 t) (at max. Q)	Color Code	Tuning Range
L-57-1	.30 MHz - 1.0 MHz	175 uh	Blue	3/1
L-57-2	1.00 MHz - 10.0 MHz	125 uh	Red	2/1
L-57-3	.01 MHz5 MHz	204 uh	Gray	3/1
L-57-6	10.00 MHz - 50.0 MHz	115 uh	Yellow	2/1
L-57-10	25.00 MHz - 100.0 MHz	100 uh	Black	2/1
L-57-17	50.00 MHz - 150.0 MHz	67 uh	Violet	1.5/1

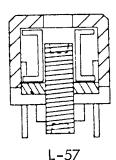




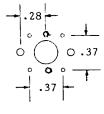








- 1. Available in materials 1, 2, 3, 6, 10, and 12
- 2. Can be tuned from both top and bottom
- 3. Furnished with six pin base to accommodate center tapped coils.



Terminal Spacing

Most efficient when tuning slug is set at maximum L. For tuning flexibility, calculate so that slug will be about 90% maximum L when at operating frequency.

Ferrite Cores and Materials

NICKEL ZINC ferrite cores exhibit high volume resistivity, moderate stability and high 'Q' factors for the 500 KHz to 100 MHz frequency range. They are well suited for low power, high inductance resonant circuits and wideband applications. Permeability range 125 mu to 850 mu.

The MANGANESE ZINC group of ferrites, have permeabilities ranging from 850 to 5000 mu. They have fairly low volume resistivity and moderate saturation flux density. High 'Q' factors for the 1 KHz to 1 MHz frequency range. Widely used for switched mode power conversion transformers operating in the 20 KHz to 100 KHz frequency range. Frequency attenuation range 20 MHz. to 400 MHz and above.

Turns formula Key to part number

Turns =
$$1000\sqrt{\frac{\text{desired L (mh)}}{\text{A}_{L} (\text{mh}/1000 +)}}$$

FT - $\frac{50}{\text{OD}}$ - $\frac{61}{\text{material}}$

MATERIAL 33 (u=850) Manganese-zinc. Low volume resistivity. Suitable for 1 KHz to 1 MHz. applications. Most commonly used for antenna rods.

MATERIAL 43 (u=850) Nickel-zinc. High volume resistivity. Widely used for medium frequency inductors and wideband transformers up to 50 MHz. Very good frequency attenuation from 30 MHz to 400 MHz. Toroids and Ferrite Beads.

MATERIAL 61 (u=125) A nickel-zinc material. Moderate temperature stability and high 'Q' for frequencies 0.2 MHz to 15 MHz. Also commonly used for wideband transformers up to 200 MHz. Toroids, Rods, and Two-hole Baluns.

MATERIAL 63 (u=40) High volume resistivity and low permeability. High 'Q' for frequency range of 15 MHz to 25 MHz. Toroidal form only.

MATERIAL 64 (u=250) A nickel-zinc material. High volume resistivity. Frequency range for resonant application up to 4 MHz. Good attenuation of unwanted frequencies up to 1000 MHz. Available in Ferrite Beads only.

MATERIAL 67 (u=40) Nickel-zinc. Very similar to the 63 material but has a greater saturation flux density. Lower volume resistivity but good temperature stability. High 'Q'applications 10 MHz to 80 Mhz. Wideband up to 200 MHz.

MATERIAL 68 (u=20) A nickel-zinc material having high volume resistivity and excellent temperature stability. High Q' resonant circuits 80 MHz.to 180 MHz, also wideband amplifiers and linear power amplifiers. Toroidal form only.

MATERIAL 72 (u=2000) Low volume resistivity. High 'Q' to 500 KHz. Very good attenuation for unwanted frequencies from 500 KHz through 50 MHz. Toroids only.

MATERIAL 73 (u=2500) Primarily a ferrite bead material. Very good attenuation properties from 0.5 MHz. through 50 MHz. Available in Ferrite Bead form only.

MATERIAL 'J'/75 (u=5000) Low volume resistivity and low core loss from 1 KHz.to 1 MHz. Pulse transformers, low level wideband transformers, and for noise attenuation from 0.5 MHz to 20 MHz. Toroids and Ferrite Beads.

MATERIAL 77 (u=2000) High saturation flux density at high temperature. Low core loss in the 1 KHz to 1 MHz range. Suited for power conversion and wideband transformers. Widely used for noise attenuation in the 0.5 MHz. to $50 \, \text{MHz}$. frequency range. Toroidal cores, Pot cores, E-cores, and Ferrite Beads.

MATERIAL 'F' (u=3000) Similar to the 77 material but has greater initial permeability. High saturation flux density at high temperature. Used for power conversion transformers. For noise attenuation in the 0.5 MHz. to 50 MHz. frequency range. Available in toroidal core form only.

FERRITE TOROIDAL CORES

Phys	sical D	imensi	ons	- Ferr	ite To	roids
core\/size	OD inches	ID inches	Hgt inches	Mean length cm	Cross Sect	Volume cm ³
FT-23 FT-37 FT-50 FT-50 -A FT-50 -B FT-82 FT-87 -A FT-114 FT-114-A FT-140 FT-150 FT-150-A FT-193-A FT-240	.230 .375 .500 .500 .825 .870 1.142 1.142 1.400 1.500 1.500 1.932 2.400	.120 .187 .281 .312 .312 .520 .540 .750 .750 .750 .750 .750 .750 1.250 1.400	.060 .125 .188 .250 .500 .250 .295 .500 .295 .500 .750 .500	1.34 2.15 3.02 3.18 3.18 5.26 5.42 7.42 7.42 9.02 8.30 8.30 12.31 14.40	.021 .076 .133 .152 .303 .246 .315 .375 .690 .806 .591 1.110 1.460 1.570	.028 .163 .402 .483 .963 1.294 1.710 2.783 5.120 7.270 4.905 9.213 18.000 22.608

A L The 63 &	To	complete	the part	number ad	ld the M	ix numbe	r to the C	ore size m	oroic number 1s respec	
Material core size	> 43	61	63	67	68	72	75	77	F	J
	u=850	u=125	u=250	u=40	u=20	u=2M	u=5M	u=2M	u=3M	u=5M
FT-23 FT-37 FT-50 FT-50 -A FT-82 FT-87 -A FT-114 FT-114-A FT-140 FT-150	188 420 523 570 1140 557 NA 603 NA 952 NA NA	24.8 55.3 68.0 75.0 150.0 73.3 NA 79.3 146.0 140.0 NA	7.9 17.7 22.0 24.0 48.0 22.4 NA 25.4 NA 45.0 NA	7.8 17.7 22.0 24.0 48.0 22.4 NA 25.4 NA 45.0 NA NA	4.0 8.8 11.0 12.0 12.0 11.7 NA 12.7 NA NA NA NA	396 884 1100 1200 2400 1170 NA 1270 2340 2350 NA	990 2210 2750 2990 NA 3020 NA 3170 NA 6736 NA	356 796 990 1080 2160 1060 NA 1140 NA 2340 NA	NA NA NA NA NA 3700 1902 NA NA 2640 5020	NA NA NA NA 3020 6040 3170 NA 6736 4400 8370
FT-193-A	NA	NA	NA	NA	NA	NA	NA	NA	4460	7435
FT-240	1240	173.0	53.0	53.0	NA	3130	6845	3130	NA	6845

Magnet	ic]	Prop	ert	ies	– F	erri	te 1	Mate	ria	Ls
Material >	43	61	63	67	68	72	75	77	F	J
Initial Perm.	850	125	40	40	20	2000	5000	2000	3000	5000
Max Perm.	3000	450	125	125	40	3500	8000	6000	4300	9500
Max Flux den. 14 oer, gauss	2750	2350	1850	3000	2000	3500	3900	4600	4700	4300
Residual flux density,gauss	1200	1200	750	1000	1000	1500	1250	1150	900	500
Vol. Resist. ohms/cm	1x10 ⁵	1x10 ⁸	1x10 ⁸	1x10 ⁷	1x10 ⁷	1x10 ²	5x10 ²	1x10 ²	1x10 ²	1x10 ²
Temp. Co-eff. 20-70 deg. C	1%	.15%	.10%	.13%	.06%	.60%	.90%	.60%	.25%	.4%
Curie Temp. C	130	350	450	500	450	150	160	200	250	140
Resonant Cir. Freq. MHz	.01 to 1 MHz	.2 to 10 MHz		10 to 80 MHz	80 to 180 MHz	.001- 2 1 MHz	.001- 1 MHz	.001- 1 MHz	.001- 1 MHz	.001- 1 MHz
Wideband Freq. MHz. *	1 to 50 MHz	10 to 200	25 to 200	50 to 500	200- 1000	.5 to 30 MH	.2 to 15 MHz	.5 to 30 MHz	.5 to 30 MHz	1 to 15 MHz
Attenuation RF Noise, MHz	20- 600	200- 1000	500- 2000	350- 1500	1000- 5000	1 - 50	·5- 20	1 - 50	1 - 50	•5 - 20

 $[\]star$ Based on low power, small core applications: Listed frequencies will be lower with high power.

FERRITE BEADS

A Ferrite bead is a dowel-like device which has a center hole and is composed of ferromagnetic material. When placed on to a current carrying conductor it will act as an RF choke. It offers a convenient, inexpensive, yet a very effective means of RF shielding, parasitic suppression and RF decoupling.

The most common noise generating suspects in high frequency circuits are power supply leads, ground leads and connections, and interstage connections. Adjacent leads and unshielded conductors can also provide a convenient path for the transfer of energy from one circuit to another. A few ferrite beads of the appropriate material placed on these leads can greatly reduce or completely eliminate the problem. Best of all, they can be added to most any existing electronic circuit.

The amount of impedance is a function of both the material and the frequency, as well as the size of the bead. As the frequency increases, the permeability will decline causing the losses to rise to a peak. With a rise in frequency the bead will present a series resistance with very little reactance. Since reactance is low there is little chance of resonance which could destroy the attenuation effect. Impedance is directly proportional to the length of the bead, therefore impedance will be additive as each similar bead is slipped onto the conductor. Since the magnetic field is totally contained within, it does not matter if the beads are touching or separated. Ferrite beads do not have to be grounded and they cannot be detuned by external magnetic fields.

We recommend the #73 or the #77 ferrite bead material for the attenuation of RFI resulting from transmissions in the amateur band. The #43 material will provide best RFI attenuation from 30 to 400 MHz, and the #64 material is most effective above 400 MHz. The #75 material is recommended for RFI from 1 to 20 MHz, but they can also be very effective even below the AM broadcast band.

Ferrite beads are usually quite small and as a result only one pass, or a small number of turns are possible. On the other hand, a toroidal core usually has a much larger ID and will accept a greater number of turns. If a large amount of impedance is required the ferrite core can be used to advantage, since the impedance increases as to the number of turns squared.

The number of turns on a single hole Ferrite bead or a toroidal core is identified by the number of times the conductor passes through the center hole. To physically complete one turn it would be necessary to cause the wires to meet on the outside of the device, however the bead or core does not care about the termination of each end of the wire and considers each pass through the center hole as one turn. (This does not apply to multihole beads)

When winding a six-hole bead, the impedance depends upon the exact winding pattern. For instance, it can be wound clock-wise or counter clock-wise progressively from hole to hole, or criss-crossed from side to side, or each turn can be completed around the outside of the bead. Each type of winding will produce very different results. The impedance figures for the six-hole bead in our chart is based on the current industry standard, which is two and one half turns threaded through the holes, criss-crossing from one side to the other.

Temperature rise above the Curie point will cause the bead to become non-magnetic, rendering it useless as a noise attenuating device. Depending on the material, Curie temperature can run anywhere from 120°C to 500°C . See 'Magnetic Properties' chart for specifics.

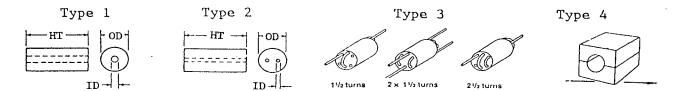
The #73 and #75 materials, as well as other very high permeability materials are semi-conductive and care should be taken not to position the cores or beads in such a manner that they would be able to short uninsulated leads together, or to ground. Other lower permeability materials with higher resistivity are non-conductive and this precaution is not necessary.

Ferrite Shielding Beads										
Part number	Bead type	Dimens OD	sions (ID	inches) Hgt	$^{ m A_L}_{ m 43}$ of	Materi 64	als (mh 73	/1000 t 75	urns) 77	Impedance factor*
FB-()-101	1	.138	.051	.128	510	150	1500	3000		1.00
FB-()-201	1	.076	.043	.150	360	110	1100			0.70
FB-()-301	1	.138	.051	.236	1020	300	3000			2.00
FB-()-801	1	.296	.094	.297	1300	390	3900			2.60
FB-(64}-901	2	.250	.050	.417		1130				7.50 ***
FB-()-1801	1	.200	.062	.437	2000	590	5900			3.90
FB-()-2401	1	.380	.197 .	.190	520		1530			1.02
FB-()-5111	3	.236	.032	.394	3540	1010				6.70 %%
FB-()-5621	1	.562	.250	1.125	3800				9600	6.40
FB-()-6301	1	.375	.194	.410	1100				2600	1.70
FB-(43)-1020	1	1.000	.500	1.112	3200					6.20
FB-(77)-1024	1	1.000	.500	.825					5600	3.70
2X-(43)-151	4	1.020	.500	1.125	Split bea	d, 43 Ma	t. Z=159	@ 25 MHz	. Z=245	5 @ 100 MHz
2X-(43)-251	4	.590	.250	1.125	Split bea					6 @ 100 MHz.

Notes: Complete the part number by adding material number in space (--) provided.

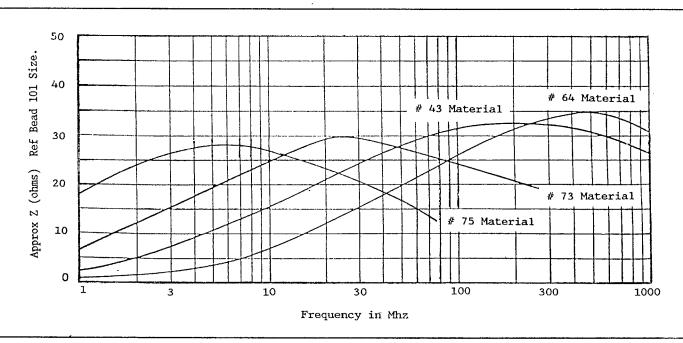
A_L values based on low frequency measurements. (mh/1000 turns) = nanohenries/turns²

*** Based on a single 'U-turn' winding. **** Based on a 2 1/2 turn, side to side winding.



Material vs Frequency vs Impedance

* Impedance Factor: This chart is based upon the `101' size bead. Impedances for other size beads may be approximated as follows: Find the `Z' of the same material at your operating frequency in the chart below. Multiply that `Z' by the Impedance Factor shown above.



Ferrites for RFI

Ferrite toroidal cores, as well as beads, can be very useful in attenuation of unwanted RF signals but we do not claim them to be a cure-all for all RFI problems. There are different types of noise sources, each of which may require a different approach. When dealing with any noise problem it is helpful to know the frequency of the interference. This is valuable when trying to determine the correct material as well as the maximum turns count.

RFI emanating from such sources as computers, flashing signs, switching devices, diathermy machines, etc. are very rich in harmonics and can create noise in the high and very high frequency regions. For this type of interference, the #43 material is probably the best choice since it has very good attenuation in the 20 MHz to 400 MHz. region. Some noise problems may require additional filtering with hi-pass or lo-pass filters. If the noise is of the differential-mode type, an AC line filter may be required. See section on AC line filters and DC chokes.

In some cases the selected core will allow only one pass of the conductor, which is considered to be one turn. In other cases it may be possible to wind several turns on to the core. When installing additional cores on the same conductor, impedance will be additive. When multiple turns are passed through a core, the impedance will increase in relation to the number of turns squared.

Keep in mind that because of the wide overlap in frequency range of the various materials, more than one material can provide acceptable results. Normally, the 43 material is recommended for frequency attenuation above 30 MHz., the 77, and 'F' materials for the amateur band, and the 'J' or materials for everything lower than the amateur band.

Computers are notorious for RF radiation, especially some of the older models which were made when RFI requirements were quite minimal. RFI can radiate from inter-connecting cables, AC power cords and even from the cabinet itself. ALL of these sources must be eliminated before complete satisfaction can be achieved. First, examine the computer cabinet to make sure that good shielding and grounding practices have been followed. If not, do what you can to correct it. If you suspect that RF is feeding back into the AC power system from your computer, wrap the power cord through an FT-240-77 toroidal core 6 to 9 times. This will act as an RF choke on the power cord and should prevent RF from feeding back into the power system where it can affect other electronic devices.

It is possible for an unwanted RF signal to enter a piece of equipment by more than one path, If so, ALL of these paths must be blocked before there will be noticeable effect. Don't overlook the fact that RFI may be entering the equipment by radiation directly from your antenna feed line due to high SWR. This, of course, can be checked with an SWR meter, and can be corrected by installing an antenna balun, or by placing a few ferrite beads, or sleeves, over the transmission line at the antenna feed point. This should prevent RF reflection back into the outside shield of the coax feed line, which could radiate RFI.

Split bars are especially designed for computer flat ribbon cables. Two or more cores can be placed on the same cable, in which case the impedance will be additive. See following page for more specific information.

RFI in telephones can be substantially reduced with the insertion of an RF choke in each side of the talk circuit. Wind two FT-50A-75 cores with about 20 turns each of #26 enamelled wire. If possible, place one in each side of the talk circuit within the telephone base. If this is not possible, try mounting them in a small box with phone modular input and output jacks mounted in each end. This can now be used 'in-line' between the phone and the wall jack. Similar results can be achieved by winding 6 to 9 turns of the telephone-to-wall cable through an FT-140-J ferrite toroidal core.

FERRITE CORES FOR RFI SUPPRESSION

Following is a list of a few of the larger Ferrite Beads (FB), Ferrite Toroidal Cores (FT), and Split Ferrite Cores (2X), all of which are extensively used for RFI problems involving multiple wire bundles, coaxial cables, microphone cables, AC power cords, telephone cables, computer cables, etc.

The 43 material is a good all around material for most RFI problems. However the lower frequencies from .5 to 10 MHz. can best be served with the 'J' or 75 material. The 77 material can provide excellent attenuation of RFI caused by amateur radio frequencies from 2 to 30 MHz. and the 43 material is best for everything above 30 MHz. however, it is still very effective across the entire amateur band but not quite as good as the 77 material. The 73 material is specifically a small ferrite bead material having a permeability of 2500 and can provide RF attenuation very similar to the 77 ferrite core material.

When a number cores are strung on the same conductor, the total impedance will be the sum impedance of all cores. When a conductor is passed through the center hole of a toroidal core a number of times, the impedance will increase in proportion to the number of turns squared.

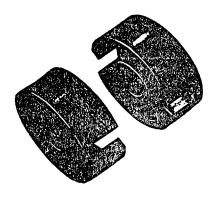
Split beads and bars (2X) are now available and can be installed without removing the end connector from the cable. Split bars are especially designed for computer ribbon cables. They are presently available for 1.3", 2.0" and 2.5" computer ribbon cables. For greater attenuation use additional cores.

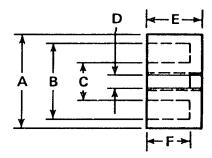
Below are a few of the most widely used cores for RFI, showing typical 'Z' in ohms for one turn at 25 MHz and 100 MHz. Most sizes are available in 43, and 77 materials. Those sizes available in the J material are marked with *

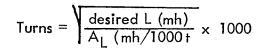
	Part number	A dim. (in)	B dim. (in)	C dim. (in)	25 MHz	100 MHz	
* * * * * * * * * * * * * * * * * * *	FT-114-43 FT-114-77	.500 .500 .870 1.142 1.142 1.400 1.400 1.930 2.400 2.400	.312 .312 .540 .750 .750 .900 .900 1.250 1.400	.250 .500 .500 .500 .295 .295 .500 .500 .625 .500	56 74 below 27 35 47 62	10 MHz 90 60 10 MHz 47 29 75 50 10 MHz 108 66	Two turns Illustrated
- A - C	2X-43-251 2X-43-151	.590 1.020	.250	1.125. 1.125	171 159	275 245	
<u>†</u> B	FB-43-1020 FB-77-1024 FB-43-5621 FB-77-5621 FB-43-6301 FB-77-6301	1.000 1.000 .562 .562 .375 .375	.500 .500 .250 .250 .194	1.120 .825 1.125 1.125 .410 .410	155 166 171 270 55 73	235 135 250 215 48 59	
A C	2X-43-651 2X-43-951 2X-43-051	for 2.	3" ribbo 0" ribbo 5" ribbo	n cable	97 105 90	200 285 250	

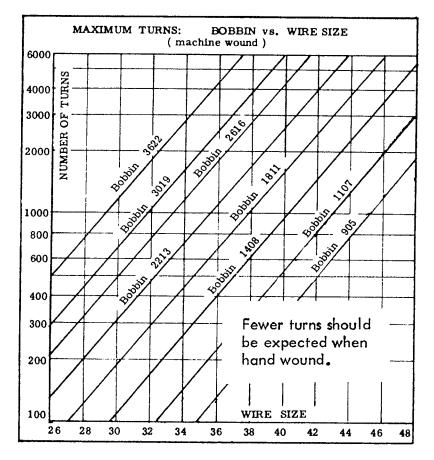
Ferrite POT Cores

Ferrite Material #77, 2000 Permeability









Physical Dimensions (In millimeters)

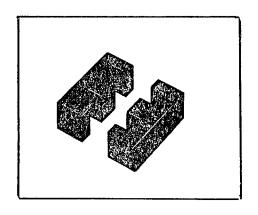
part mumber	A	В	С	D	E	F
PC-1107-77	11.10	9.20	4.60	2.10	3.21	2.27
PC-1408-77	14.05	11.80	5.90	3.10	4.18	2.90
PC-1811-77	18.00	15.25	7.45	3.10	5.27	3.70
PC-2213-77	21.60	18.70	9.25	4.55	6.70	4.70
PC-2616-77	25.50	21.60	11.30	5.55	8.05	5.60
PC-3019-77	30.00	25,40	13.30	5.55	9.40	6.60
PC-3622-77	35.60	30.40	15.90	5.55	10.85	7.40

Magnetic Dimensions

part number	$^{\mathrm{A}}\mathrm{e}$	$1_{\mathbf{e}}$	$v_{\mathbf{e}}$	$\mathtt{A}_{\mathtt{L}}$	Power
	mm^2	mm	mm ³	mh/1000-t	Based on 20 KHz
PC-1107-77	15.9	15.9	252	1420	Max 3 watts
PC-1408 77	25.0	20.0	500	1960	Max 5 watts
PC-1811-77	43.0	25.9	1120	2880	Max 10 watts
PC-2213-77	63.0	31.6	2000	3660	Max 20 watts
PC-2616-77	93.0	37.2	3460	4700	Max 50 watts
PC-3019-77	136.0	45.0	6100	5900	Max 70 watts
PC-3622-77	202.0	53.0	10600	7680	Max 90 watts

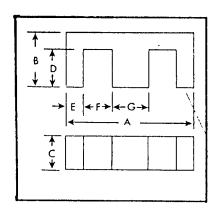
Note: Power ratings are conserative, based on 20 KHz. switching frequency.

FERRITE 'E' CORES



TYPE 77 FERRITE MATERIAL permeability 2000

These cores are ideally suited for low power applications up to 200 watts. A nylon bobbin is supplied for easy winding.



E-Core	Physical	Dimensions	(inches)

Part No.	A	В	С	D	E	F	G	Power
EA-77-188	.760	.318	.187	.225	.093	.192	.187	10 watts
EA-77-250	1.000	.380	.250	.255	.125	.250	.250	20 watts
EA-77-375	1.375	.562	.375	.375	.187	.312	.375	70 watts
EA-77-500	1.625	.650	.500	.405	.250	.312	.500	100 watts
EA-77-625	1.680	.825	.605	.593	.234	.375	.468	200 watts

E-Core Magnetic Properties

Part No.	A _e mm ²	I _e mm	Ve mm³	A _S mm ²	$rac{A_{ m W}}{ m mm}^2$	A _C x A _w	A _L value mh/1000 turns
E-77-188	22.5	40.1	900	1050	55.7	1250	1290
E-77-250	40.4	48.0	1930	1700	80.6	3250	1520
E-77-375	90.3	68.8	6240	3630	151.0	13700	2540
E-77-500	160.0	76.7	12300	5410	163.0	26100	4090
E-77-625	184.0	98.0	18000	7550	287.0	52900	5210

	Wire	siz	:e	vs.	N	lumb	er	of	tur	ns	
Part No.	18	20	22	24	26	28	30	32	34	36	38
EA-77-188	21	33	50	79	125	196	293	439	669	1046	1548
EA-77-250	34	62	93	147	232	364	532	814	1240	1938	
EA-77-375	63	94	149	235	372	582	868	1302	1984		
EA-77-500	50	141	212	335	532	829	1236	1855			
EA-77-625	159	250	375	593	939	1470	2191				

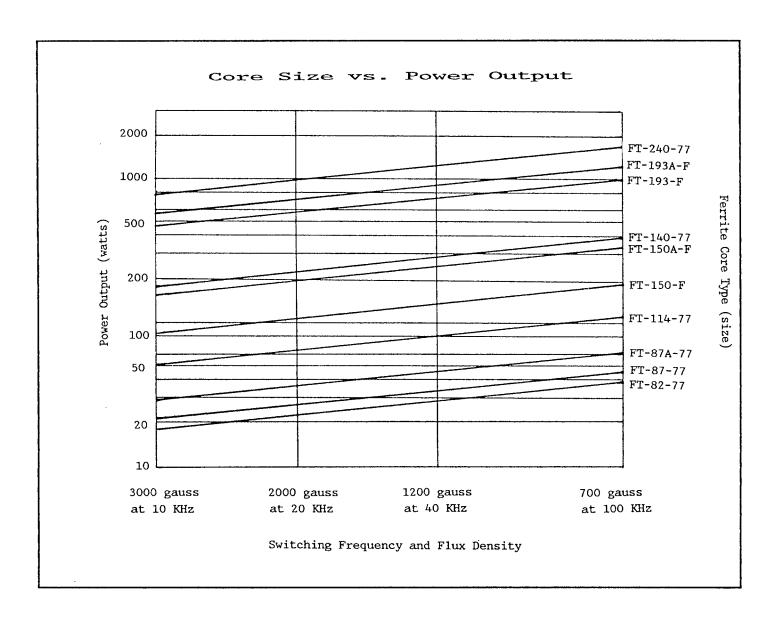
Switched Mode Power Supplies

Guide to select the proper size ferrite core

Switching power supplies require the use of high permeability Ferrite type cores, rather than high permeability Iron Powder cores. High permeability Iron Powder cores will be to lossy.

Toroidal cores may be used, however 'E' type cores are generally preferred because of greater winding ease. We stock both the Toroidal Ferrite cores and the 'E' cores in the #77 material, which is ideal for switching at frequencies of 20 KHz or higher.

See other pages in this brochure on 'E' cores for size vs. power. The chart at the bottom of this page will provide data on an approximate size toroidal core to be used for a given amount of power.



BALUNS and WIDEBAND CORES

The two-hole balun is commonly used for wideband transformers and impedance matching devices. The primary concern, when designing a wideband transformer, is to extend the banddwidth with a minimum of loss. The limiting factors are inductive reactance and core loss.

By winding through both holes of the binocular type two hole balun, a higher inductance per turn can be obtained than would otherwise be possible with a single hole core.

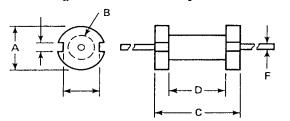


Dimensions in inches; A_T value in mh/1000 turns based on hole to hole winding

Part No.	OD	ID	Hgt	Th	Type	^{A}L	Part No.	OD	ID	Hgt	Th	Type	$^{A}_{\mathtt{L}}$
BN-43-202	.525	.150	.550	.295	one	2890	BN-61-2302	.136	.035	.093	.080	one	100
BN-43-2302	.136	.035	.093	.080	one	680	BN-61-2402	.280	.070	.240	.160	one	280
BN-43-2402	.280	.070	.240	.160	one	1277	BN-61-1702	.250	.050	.470		two	420
BN-43-3312	.765	.187	1.000	.375	one	5400	BN-61-1802	.250	.050	.240		two	310
BN-43-7051	1.130	.250	1.130	.560	one	6000	BN-73-202	.525	.150	.550	.295	one	8500
BN-61-202	.525	.150	.550	.295	one	425	BN-73-2402	.275	.070	.240	.160	one	3750

Ferrite Bobbin Cores

Ferrite bobbins provide a convenient means of winding RF chokes. Because of their open magnetic path, they can handle more current than toroids of similar size. To aid in the design of such chokes, we have provided $A_{\rm L}$ values, a winding table, and ampere-turn ratings for each bobbin.



Winding table: number of turns to completely fill bobbin.

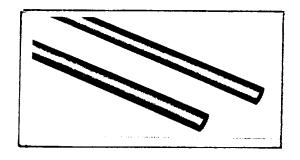
22 24 26 28 30 32 34 36 wire size B-72-1111 14 23 35 56 88 205 400 164 20 22 24 26 28 32 wire size 30 34 B-72-1011 24 39 60 93 148 230 425

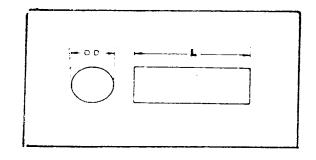
BOBBIN	DIMENSIC	NS			A _L valu	ein mh/1000	turns
part number	Α	В	C	D	F	Αı	1

part number	А	В	C	Ü	۲	Αl	NI
Bobbin # B-72-1111	.196"	.107"	.500"	.400"	#22	1 <i>7</i>	60
Bobbin # B-72-1011	.372"	.187"	.750"	.500"	#20	39	130

BOBBIN # B-72-1111		$A_L = 17$	NI= 60	I= 60 BOBBIN # B-72-1011		-72-1011	$A_L = 39$ NI = 130	
Inductance	wire turns	wire size	l (max)	,	Inductance	wire turns	wire size	l (max)
10 uh	24	24	2.50		25 uh	2 5	20	5.20
25 uh	38	26	1.60	ļ	50 uh	36	22	3.60
50 uh	54	28	1.10	ļ	100 uh	50	24	2.60
100 uh	77	30	.78	_ _	250 uh	80	26	1.60
250 uh	121	31	.50		500 uh	113	27	1.10
500 uh	171	32	.35		1.0 mh	160	2 8	.80
1.0 mh	243	34	.25	_	2.5 mh	253	30	.50
2,5 mh	383	36	.16	į	5.0 mh	358	32	.36
5.0 mh	542	3 <i>7</i>	.11		10.0 mh	506	34	.25
10.0 mh	762	38	.08		25.0 mh	800	36	.16

Ferrite Rods





Part number	Material	Perme- ability	Diameter (in)	Length (in)	Al value mh/1000 t	Ampere turns
R61-025-400	61	125	.25	4.0	26	110
R61-033-400	61	125	.50	4.0	32	185
R61-050-400	61	125	.50	4.0	43	575
R61-050-750	61	125	.50	7.5	49	260
R33-037-400	33	800	.37	4.0	62	290
R33-050-200	33	800	.50	2.0	51	465
R33-050-400	33	800	.50	4.0	59	300
R33-050-750	33	800	.50	7.5	70	200

FERRITE RODS are available in various sizes of both the #33 and #61 materials, which are standard stock items here at Amidon. The most common use of a ferrite rods is for antennas and choke applications.

ANTENNAS: The #61 material rods are widely used for commercial AM radio antennas and on up to 10 MHz. The #33 material rods are more suitable for the VLF frequency range.

CHOKE APPLICATIONS; Both the #33 and the #61 material rods are extensively used in choke applications. The #33 material should be selected for the 40 and 80 meter bands and the #61 material is most suitable for 10 through 40 meters. The #33 material rods are also often used in speaker cross-over networks. Due to the open magnetic structure of the rod configuration, considerable current can be tolerated before it will saturate.

There are several factors that have a direct bearing on the effective permeability of a ferrite rod, which in turn will effect inductance and ${}^{{}^{{}}}Q{}^{{}^{{}}}$, as well as the A_L value of the rod and its ampere-turns rating. These are: (1) Length to diameter ratio of the rod, (2) Placement of the coil on the rod, (3) Spacing between turns and, (4) Air space between the coil and the rod. In some cases the effective permeability of the rod will be influenced more by a change in the length to diameter ratio than by a change in the initial permeability of the rod. At other times, just the reverse will be true.

Greatest inductance and ${\rm A_L}$ value will be obtained when the winding is centered on the rod, rather than placed at either end. The best 'Q' will be obtained when the winding covers the entire length of the rod.

Because of all of the above various conditions it is very difficult to provide workable $\ A_L$ values, however we have attempted to provide a set of A_L and NI values for various types of rods in our stock. These figures are based on a closely wound coil of #22 wire, placed in the center of the rod and covering nearly the entire length. Keep in mind that there are many variables and that the inductance will vary according to winding technique.